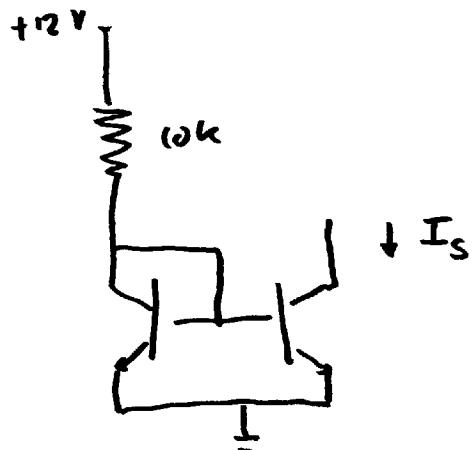
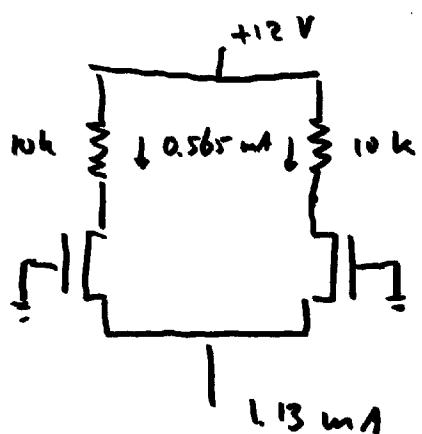


Electrónica II Exame Época Especial
5/12/2006



Fonte de corrente, I_S :

$$\frac{(12 - 0.7)V}{10k\Omega} = 1.13 \text{ mA}$$



$$I_{M1} = I_{M2} = 1.13 \text{ mA} / 2 = 0.565 \text{ mA}$$

$$V_D = 12V - 0.565 \text{ mA} \cdot 10k\Omega$$

$$= 6.35 \text{ V} = V_D$$

$$V_G = 0$$

SAT or LIN?

$$\text{SAT: } I_D = \frac{k}{2} |V_{GS} - V_T|^2$$

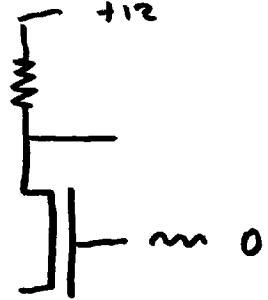
$$565 \mu\text{A} = 100 \frac{\mu\text{A}}{\text{V}^2} \cdot (V_{GS} - 1)^2 \Rightarrow V_{GS} = 3.38 \text{ V}$$

$$\Rightarrow V_S = -3.38 \text{ V} \quad (V_G = 0)$$

$$V_D = 6.35 \text{ V}$$

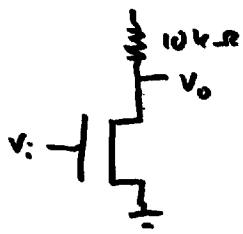
$$V_G = 0$$

$$\left. \begin{aligned} V_{DS} &= 6.35 - (-3.38) = 9.73 \text{ V} \\ V_{GS} &= 3.38 \text{ V} \end{aligned} \right\} V_{DS} > V_{GS} \Rightarrow \text{SAT}$$



Esta base polarizada. O transistor fica
mais ou menos no meio de 0
(V_G) e +12 V (V_{CC})

b)



O ganho de um source common é
 $-g_m \cdot R_D$

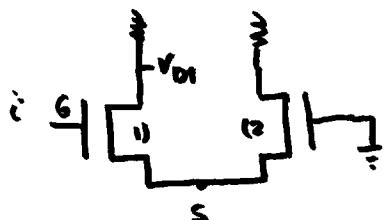
$$g_m = \frac{dI_D}{dV_{GS}} = \frac{d \frac{k}{2} (V_{GS} - V_T)^2}{dV_{GS}} = k(V_{GS} - V_T)$$

$$= 200 \frac{\mu A}{V^2} \cdot (3.38 V - 1 V) = 476 \frac{\mu A}{V}$$

$$\frac{V_D}{V_G} = -g_m R_D = -476 \frac{\mu A}{V} \cdot 10 k\Omega$$

$$= -4.76 \frac{V}{V}$$

no por diferencial



$$\frac{dV_{GS1}}{dV_{i1}} = \frac{1}{2} \quad \left. \begin{array}{l} \text{subida de } V_i \text{ é} \\ \text{dividida sobre os} \\ \text{dois transistores} \end{array} \right)$$

O ganho do p.d.

$$A_{v1} = \frac{dV_{D1}}{dV_{i1}} = -R_D \cdot \frac{dI_{D1}}{dV_{i1}} = -R_D \frac{dI_{D1}}{dV_{GS1}} \cdot \frac{dV_{GS1}}{dV_{i1}}$$

$$= -R_D \cdot 476 \frac{\mu A}{V} \cdot \frac{1}{2} = -2.38 \frac{V}{V}$$

pela simetria: $A_{v2} = +2.38 \frac{V}{V}$ ($= \frac{dV_{D2}}{dV_{i1}}$)

$$A_{dm} = |A_{v1} - A_{v2}| \approx 4.76 \frac{V}{V}$$

modo comum: $A_{cm} = 0$. Desde não há corrente do gate, para qualquer tensão de entrada igual as duas entradas

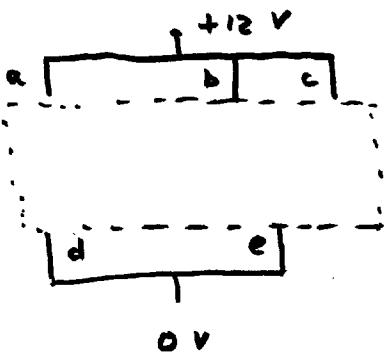
$$I_{D1} = I_{D2} = 565 \mu A$$

$$\frac{dI_{D1}}{dv_i} = \frac{dI_{D2}}{dv_i} = 0$$

$$A_V = -R_D \cdot \frac{dI_{D1}}{dv_i} = 0$$

$$CMRR = \infty$$

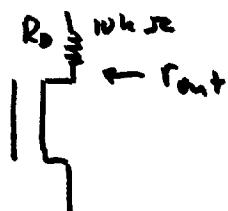
c)



$$\begin{aligned} a &= 1.13 \text{ mA} \\ b &= 0.565 \text{ mA} \\ c &= 0.565 \text{ mA} \\ d &= 1.13 \text{ mA} \\ e &= 1.13 \text{ mA} \end{aligned}$$

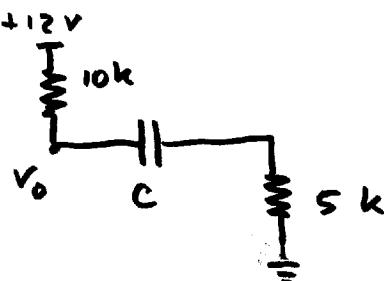
$$\begin{aligned} P &= I \cdot V = 12 \cdot (I_a + I_b + I_c) + 0 = (I_d + I_e) \\ &= 27 \text{ mW} \end{aligned}$$

d) $r_{in} = \infty$ ($I_G = 0$)



$$r_o = \left| \frac{\partial I_o}{\partial v_o} \right| = \left| \frac{d \frac{k}{2}(v_{GS} - v_t)}{\partial v_{DS}} \right| = \infty$$

e) baixas freqüências

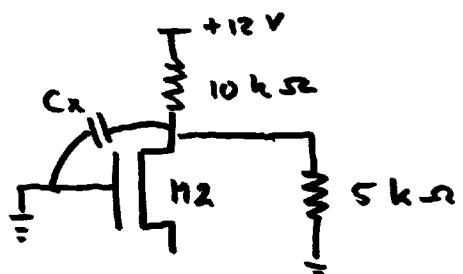


$$f_{CL} = \frac{1}{2\pi R_C} = 10 \text{ Hz}$$

(4)

$$C = \frac{1}{2\pi (10 \text{ k}\Omega // 5 \text{ k}\Omega) \cdot 10 \text{ Hz}} = 4.8 \mu\text{F}$$

altas freqüências

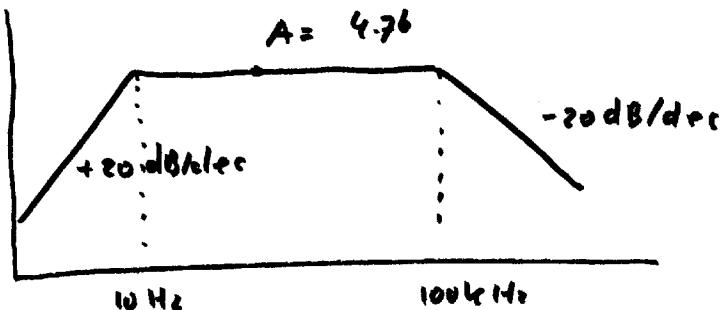


$$C_M = (1-A)C_x = (1-0)C_x = C_x$$

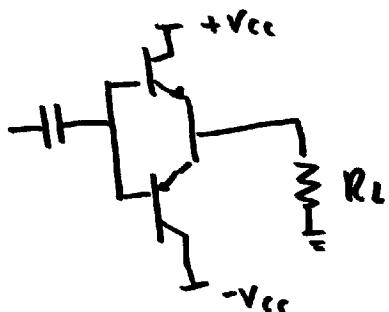
$$f_{CH} = \frac{1}{2\pi R_C} = 100 \text{ kHz}$$

$$\Rightarrow C = 48 \text{ mF}$$

f) $20 \times \log(A)$



g)



$$P_L = \frac{1}{2} \frac{V_{CC}^2}{R_L} = 1 \text{ W}$$

$$\Rightarrow V_{CC} = \sqrt{2 \cdot 8 \cdot 1 \text{ W}} = 4 \text{ V}$$

h)

$$P_S = \frac{1}{T} \int_0^T V \cdot I dt = \frac{1}{T} \int_0^T V_{CC}^2 \cdot \frac{1}{R_L} \sin \omega t dt = \frac{4}{\pi} \frac{V_{CC}^2}{R_L} = 2.5 \text{ W}$$

nos transistores

(5)

$$P_T = P_S - P_L = 1.5 \text{ W}$$

por transistor

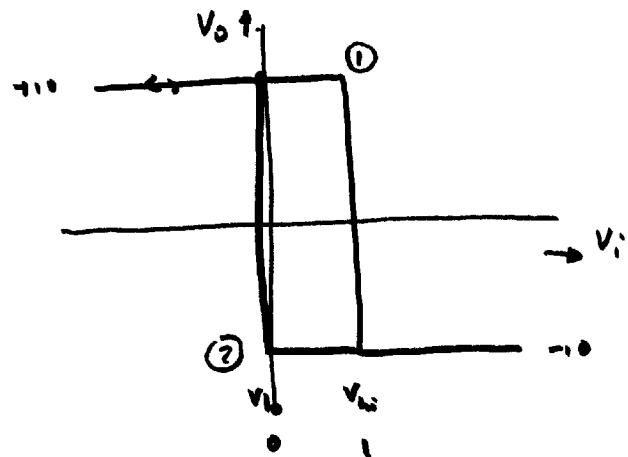
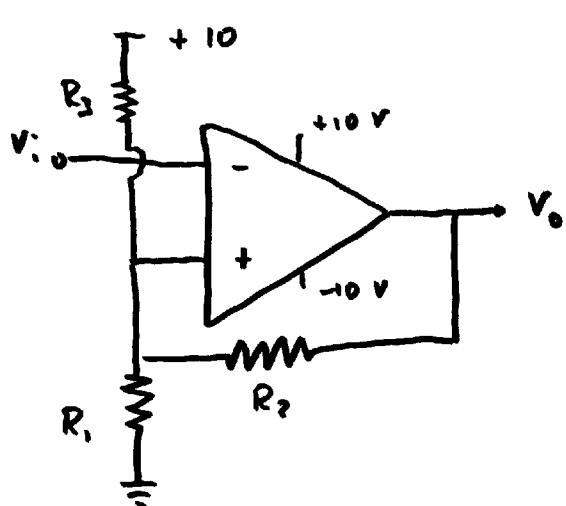
$$P_{T1} = P_{T2} = \frac{1}{2} P_T = 0.75 \text{ W}$$

$$T_J = T_A + P \cdot R_T$$

$$= 25^\circ + 0.75 \text{ W} \cdot 100^\circ\text{C/W} = 100^\circ\text{C}$$

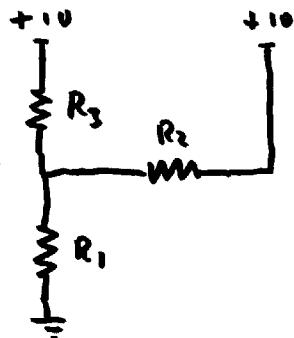
não há problema

2



① $V_o = +10 \text{ V}$

$$V_P = \frac{R_1}{R_1 + R_2//R_3} \cdot 10 \text{ V}$$

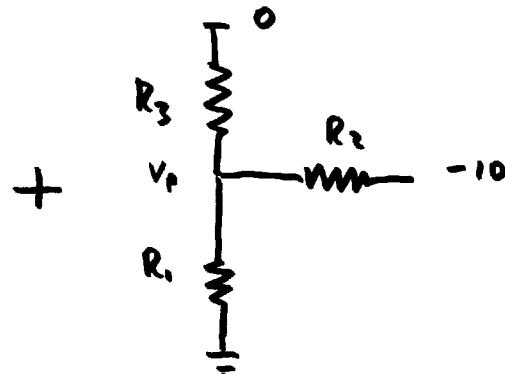
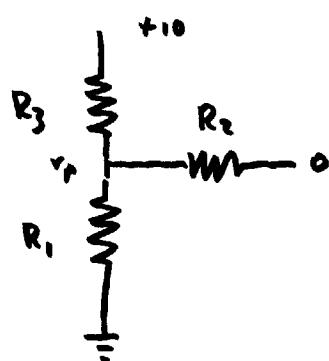
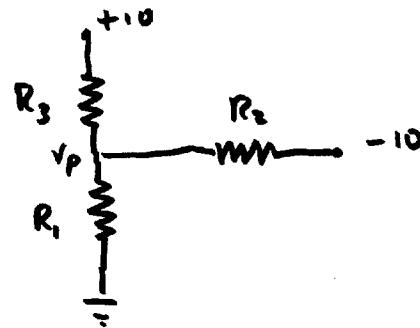


no $V_i = V_{hi}$, $V_P = V_n \Rightarrow 10 \cdot \frac{R_1}{R_1 + R_2//R_3} = 1 \quad (\text{I})$

(6)

②

$$V_0 = -10 \text{ V}$$



$$V_p = \frac{R_1 // R_2}{R_1 // R_2 + R_3} \cdot 10 + \frac{R_1 // R_3}{R_1 // R_3 + R_2} (-10) = 0 \quad (\text{II})$$

$\uparrow V_{10}$

$$\text{Erstelle : } R_1 = 1 \text{ k}\Omega$$

$$(\text{Ia}) : \frac{1 \text{ k}\Omega}{1 \text{ k}\Omega + R_2 // R_3} = \frac{1}{10} \Rightarrow \underline{\underline{R_2 // R_3 = 9 \text{ k}\Omega}}$$

$$(\text{IIa}) \quad \frac{\frac{R_2}{1+R_2}}{\frac{R_2}{1+R_2} + R_3} = \frac{\frac{R_3}{1+R_3}}{\frac{R_3}{1+R_3} + R_2}$$

$$R_2 = R_3 = 18 \text{ k}\Omega$$

$$\frac{R_2}{R_2 + R_3 + R_2 R_3} = \frac{R_3}{R_2 + R_3 + R_2 R_3}$$

$$R_2 = R_3$$