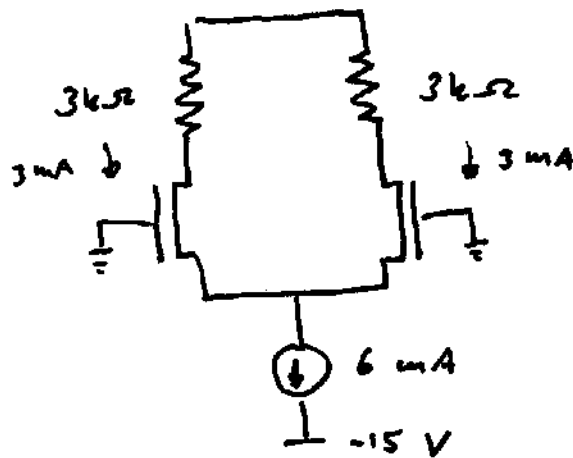


1))
a)

$$I_{D1} = I_{D2} = \frac{I_{\text{source}}}{2} = 3 \text{ mA}$$

$$V_{O1} = V_{O2} = 15 \text{ V} - 3 \text{ mA} \times 3 \text{ k}\Omega = 6 \text{ V}$$

$$V_{DS} \stackrel{?}{>} V_{GS} - V_T$$

$$V_T = 0, V_G = 0, V_D = 6 \Rightarrow$$

yes. saturation

$$I_D = \frac{k}{2} (V_{GS})^2 = 3 \text{ mA} \Rightarrow V_{GS} = \sqrt{\frac{2 \times 3 \text{ mA}}{k}}$$

$$g_m = \frac{\partial I_D}{\partial V_G} = k V_{GS} = 3.4 \text{ mS}'$$

$$\Rightarrow k = \frac{3.4 \text{ mS}'}{V_{GS}}$$

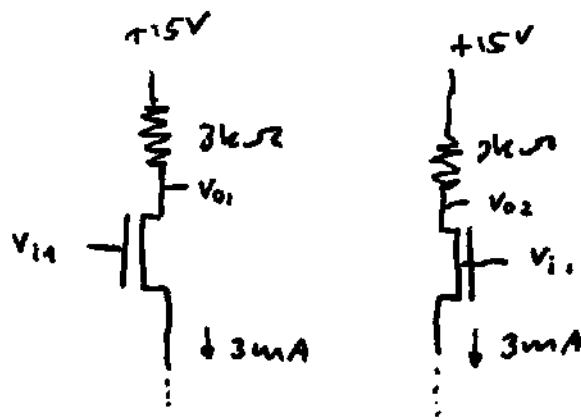
$$\Rightarrow k = \frac{3.4 \text{ mS}'}{\sqrt{6 \text{ mA}}} \times \sqrt{k} \Rightarrow \sqrt{k} = \frac{3.4 \text{ mS}'}{\sqrt{6 \text{ mA}}}$$

$$k = \frac{(3.4 \text{ mS}')^2}{6 \text{ mA}} = 1.93 \frac{\text{mA}}{\text{V}^2}$$

$$V_{GS} = \sqrt{\frac{6 \text{ mA}}{1.93 \text{ mA/V}^2}} = 1.76 \text{ V}$$

$$V_G = 0 \Rightarrow V_S = -1.76 \text{ V}$$

b) common mode gain: $A_{cm} = 0$ because of symmetry and the fact that $I_G = 0$ (therefore $I_D = I_S$, always!):



$v_{o1} = 6 \text{ V}$, always, independent of v_i

$v_{o2} = 6 \text{ V}$, always, independent of v_i

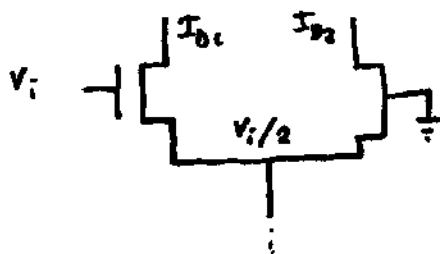
$v_{out} = v_{o2} - v_{o1} = 0$, always.

$$A_v = \frac{v_o}{v_i} = \frac{dv_{out}}{dv_i} = 0 \quad \text{common mode.}$$

Differential-mode gain: A_{dm}

For small signals, the circuit and transistors work linearly. The input v_i thus divides equally

between the two transistors.



$$\frac{dI_{D1}}{dv_i} = \frac{1}{2} \frac{dI_{D1}}{dV_{GS1}} = \frac{1}{2} k V_{GS} = \frac{1}{2} 1.93 \frac{\text{mA}}{\text{V}} = 1.76 \text{ V} = 1.7 \frac{\text{mA}}{\text{V}}$$

$$\frac{dv_{o1}}{dv_i} = -R \times \frac{dI_{D1}}{dv_i} = 3 \text{ k}\Omega \times 1.7 \frac{\text{mA}}{\text{V}} = -5.1 \frac{\text{V}}{\text{V}}$$

$$\frac{dI_{D2}}{dv_i} = -\frac{1}{2} \frac{dI_{D1}}{dV_{GS1}}, \quad \frac{dv_{o2}}{dv_i} = -R \frac{dI_{D2}}{dv_i} = +5.1 \frac{\text{V}}{\text{V}}$$

$$A_{dm} = 5.1 \text{ V/V} - (-5.1 \text{ V/V}) = 10.2 \text{ V/V}$$

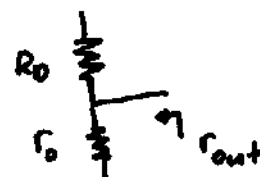
$$CMRR = \left| \frac{A_{dm}}{A_{cm}} \right| = \frac{10.2}{0} = \infty$$

$$c) \quad r_{in} = 1 / \frac{\partial I_{in}}{\partial V_{in}} = 1 / \frac{\partial I_G}{\partial V_G} = 1/0 = \infty$$

$$r_{out} = R_D \parallel r_o$$

$$r_o = V_A / I_D = \frac{100 \text{ V}}{3 \text{ mA}} =$$

$$= 33 \text{ k}\Omega$$



$$r_{out} = 3 \text{ k}\Omega \parallel 33 \text{ k}\Omega = 2.75 \text{ k}\Omega$$

2)) see lecture notes

3)) a)



$$\text{C.S. : } r_{in} = R_{G2} \parallel R_{G1} = 330 \text{ k}\Omega \parallel 1.5 \text{ M}\Omega = 270 \text{ k}\Omega$$

$$r_{out} = R_D = 2 \text{ k}\Omega$$

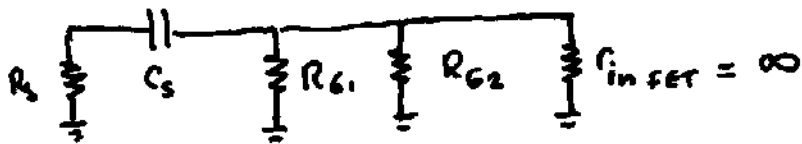
$$A_v = \frac{-R_D}{(1/g_m + R_S)} = \frac{2 \text{ k}\Omega}{294 \Omega + 820 \Omega} = -1.80$$

$$A_{total} = \frac{r_{in}}{R_{S0} + r_{in}} \times A_v \times \frac{R_L}{R_{out} + R_L} = -1.59$$

b) using short-circuit time constants and open-circuit time constants we will find time constants for each capacitance and then combine them to give f_L and f_H , the cut-off frequencies at lower and higher frequencies

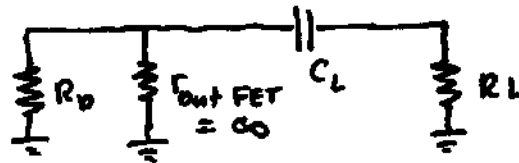
(4)

C_s : is an HPF (high-pass filter)



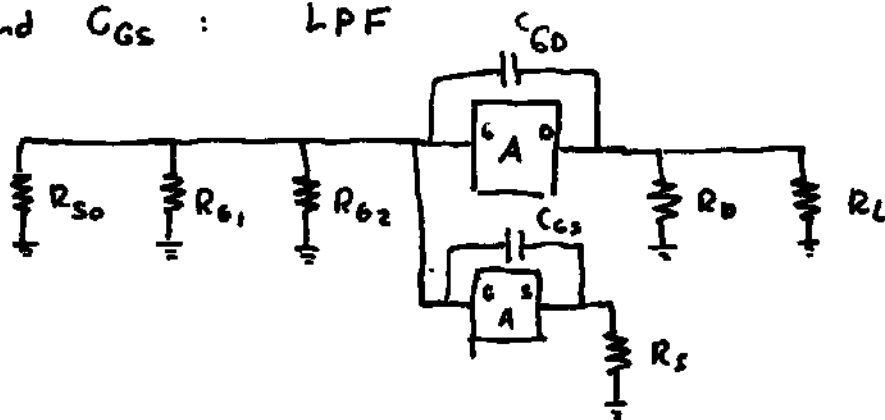
$$\tau_s = C_s (R_s + R_{G1} // R_{G2}) = 1 \mu F \times (20 k\Omega + 330 k\Omega // 1.5 M\Omega) = 0.29 s$$

C_L : is an HPF

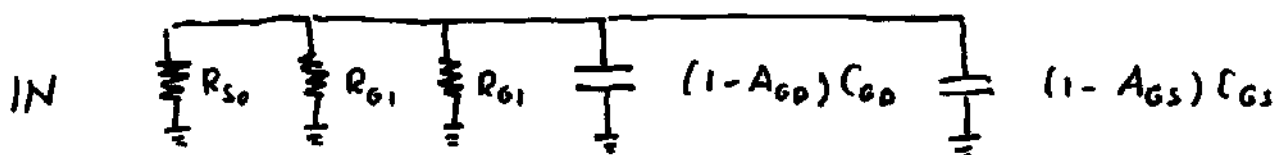


$$\tau_L = C_L (R_L + R_D) = 20 nF \times (2 k\Omega + 40 k\Omega) = 8.4 \times 10^{-4} s$$

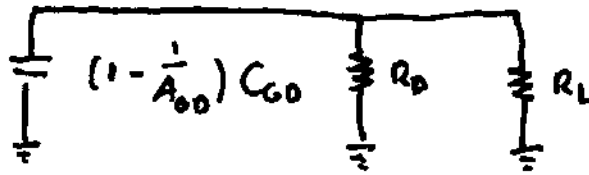
C_{GD} and C_{GS} : LPF



$$A_{GD} = -1.80, \quad A_{GS} = + \frac{R_s}{r_{gs} + R_s} = 0.74$$



OUT

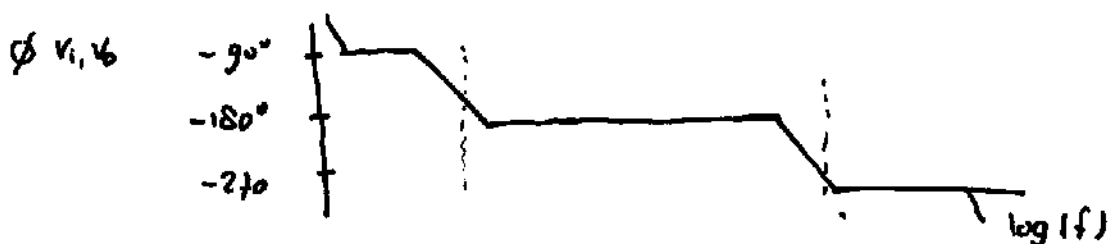
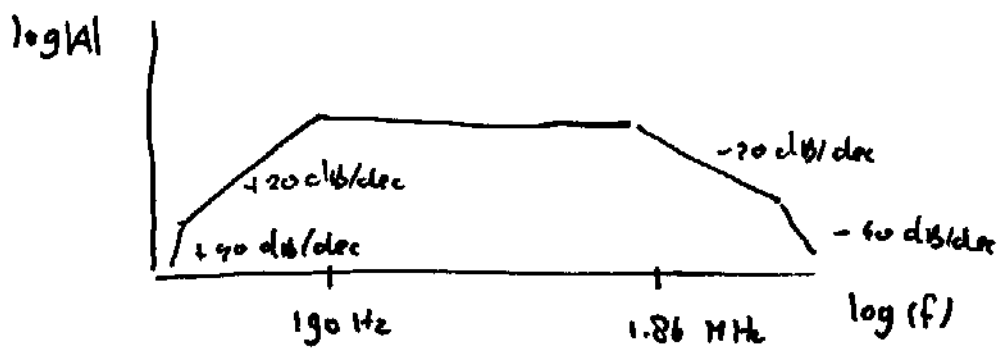


$$\begin{aligned}\tau_{in} &= \left[(1 - A_{00}) C_{00} + (1 - A_{05}) C_{05} \right] \times (R_{S0} \parallel R_{G1} \parallel R_{G1}) \\ &= [2.8 \times 1.2 \text{ pF} + 0.26 \times 4 \text{ pF}] \times 18.62 \text{ k}\Omega \\ &= 8.19 \times 10^{-8} \text{ s}\end{aligned}$$

$$\begin{aligned}\tau_{out} &= \left[(1 - \frac{1}{A_{00}}) C_{00} \right] \times [R_D \parallel R_L] \\ &= 1.56 \times 1.2 \text{ pF} \times 1.90 \text{ k}\Omega \\ &= 3.56 \times 10^{-9} \text{ s}\end{aligned}$$

$$\begin{aligned}\tau_{lo} &= \left(\frac{1}{\tau_s} + \frac{1}{\tau_L} \right)^{-1} = 8.4 \times 10^{-4} \text{ s} \\ \Rightarrow f_{lo} &= \frac{1}{2\pi\tau_{lo}} = 190 \text{ Hz}\end{aligned}$$

$$\begin{aligned}\tau_{hi} &= \tau_{in} + \tau_{out} = 8.55 \times 10^{-8} \text{ s} \\ \Rightarrow f_{hi} &= 1.86 \text{ MHz}\end{aligned}$$

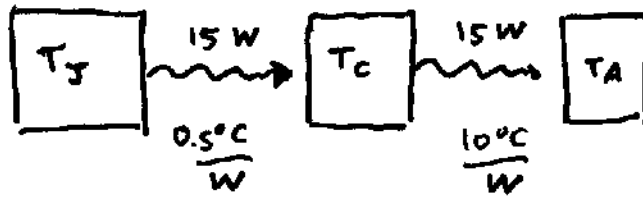


c) $190 \text{ Hz} - 1.86 \text{ MHz} \sim 1.86 \text{ MHz}$

4)) see lecture notes

5))

a)



steady-state situation

Given the fact that $T_A = 40^\circ\text{C}$,

$$T_c = 40^\circ\text{C} + 15\text{W} \times \frac{10^\circ\text{C}}{\text{W}} = 190^\circ\text{C}$$

$$T_j = T_c + 15\text{W} \times \frac{0.5^\circ\text{C}}{\text{W}} = 197.5^\circ\text{C} \Rightarrow \text{too hot!}$$

b) A dissipater lowers the thermal resistance from T_c to T_A . Reasoning other way around:

$$T_j = 150^\circ\text{C} \Rightarrow$$

$$T_c = T_j - 15\text{W} \times \frac{0.5^\circ\text{C}}{\text{W}} = 142.5^\circ\text{C}$$

$$T_A = T_c - 15\text{W} \times R_T = 40^\circ\text{C}$$

$$142^\circ\text{C} - 15\text{W} \times R_T = 40^\circ\text{C}$$

$$\Rightarrow R_T = \frac{102.5^\circ\text{C}}{15\text{W}} = 6.8^\circ\text{C/W}$$

